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Super-Twisting Algorithm-Based Virtual Synchronous Generator in Inverter Interfaced Distributed Generation (IIDG)

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Abstract: The significant proliferation of renewable resources, primarily inverter interfaced distributed generation (IIDG) in the utility grid, leads to a dearth of overall inertia. Subsequently, the system illustrates more frequency nadir and a steeper frequency response. This may degrade the dynamic frequency stability of the overall system. Further, virtual inertia has been synthetically developed in IIDG, which is known as a virtual synchronous generator (VSG). In this work, a novel STO-STC-based controller has been developed, which offers flexible inertia following system disturbance. The controller is based on the super-twisting algorithm (STA), which is a further advancement in the conventional sliding mode control (SMC), and has been incorporated in the control loop of the VSG. In this scheme, two steps have been implemented, where the first one is to categorize all states of the system using a super-twisting observer (STO) and further, it is required to converge essential states very quickly, exploiting a super-twisting controller (STC). Thus, the STO-STC controller reveals a finite-time convergence to the numerous frequency disturbances, based on various case studies. The performance of the controller has been examined in the MATLAB environment with time-domain results that corroborate the satisfactory performance of the STO-STC scheme and that illustrate eminence over the state of the art.

Keywords: virtual inertia emulation; virtual synchronous generator (VSG); inverter interfaced distributed generation (IIDG); sliding mode control (SMC); super-twisting algorithm (STALG); super-twisting control (STC)



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1. Introduction

The increasing penetration of renewable energy sources, mainly IIDG, has endangered frequency stability concerns in the low inertia system, due to the unavailability of the rotating mass [1]. Thus, inertia emulation is urgently required in IIDG to assist with frequency regulation [2]. Commonly, synchronous machine (SM) droop characteristics have been developed virtually in IIDG, which is popularly known as a virtual synchronous generator (VSG). Therefore, VSG-IIDG is essentially required in frequency regulation. A VSG control scheme has been presented in [3,4], in which the swing equation of an actual SM has been utilized. Moreover, in recent years, several VSG control schemes were presented, in which the active power extracted from IIDG has been exploited to regulate the frequency deviation, the rate of change of frequency (RoCoF), and the settling time. Nevertheless, IIDG has some active power constraints that mainly depend upon maximum power point tracking (MPPT). Further, various control schemes have been developed that are based on droop control [5,6], in which the droop characteristics are employed to regulate a fraction of the active power in accordance with the frequency deviation. In addition, the various parameters (inertia constant, damping constant, and droop slope) required to regulate the active power supplied through IIDG have been optimized through a number of presented

schemes [7,8]. Nonetheless, for simplicity, there has been barely two parameters (inertia and damping) that have been incorporated in the basic design of VSG. In addition, there is an inherent trade-off between a selection of various parameters and the accomplishment of the required aspect viz. RoCoF, the settling time, the frequency regulation, and the minimization of the frequency nadir. Further, there will be no alternative approach to regulating all of the discussed aspects simultaneously, depending upon the selection of the parameters. Lastly, it is quite problematic to tune the various parameters of VSG to achieve all of the requirements at the same time, since a variation to any of them can negatively affect another performance.

1.1. Motivation of Work

In recent years, VSG control schemes have been extensively studied and investigated, especially for IIDG, considering both the grid-connected and islanded modes. VSG-IIDG has been designed to support the frequency dynamics, and to minimize the output power and frequency oscillations. Several control techniques have been developed to suppress the frequency and power oscillations [9]. The proper selection of the inertia and damping constants can effectively reduce both of these oscillations. Further, to have a proper acquaintance with VSG design, small-signal modeling has been discussed in several papers [8,10] that provide more precise parameters for the emulated inertia. Further, a VSG based on an adaptive linear quadratic regulator has been proposed in [11]. A VSG scheme discussed in the literature has considered only a few IIDGs in the proposed test system, which lacks authenticity; thus, reliability is still an open problem. Secondly, the effectiveness of previously presented schemes in both standalone and grid-connected modes still requires intense study. The grid-connected IIDG has been recognized to be more sensitive to voltage sag/swell and exposure to unbalanced conditions. The VSG design should be well-operated in both the modes, and lastly, the controller performance is required to be more robust, as per the variation of various parameters.

1.2. Literature Survey on Recent State of Art

A self-tuning algorithm (STALG) in VSG has been proposed in [2]. Further work in self-tuning was given in [12], which was developed based on a RoCoF, and it provides the optimal virtual inertia (VI) via the proper selection of the inertia and damping constants. Since IIDG is a non-linear system, owing to the power–voltage characteristics, many control schemes were therefore developed based on the linearized model with proportional-integral viz. PI controller for inertia emulation. Subsequently, linearized feedback non-linear control schemes have been discussed in the literature, which provide enhanced performance over linear control design. Nonetheless, the non-linear design performs well on optimal parameter selection and an accurate operating point, which is slightly difficult to design. Furthermore, an AI-based VSG has been presented in [13], and further, a neural-fuzzy VSG (N-F-VSG) was designed in [14], which is a more reliable structure for enhancing the frequency dynamics. Recently, a model predictive control (MPC) has been portrayed in the literature, which has been proven to be an excellent VSG design [15,16]. The fuzzy-based VSG (Fuzzy-VSG), which computes the correction factor required to alter the governor output during a sudden disturbance, was discussed in [17]. The improved MPC scheme has been suggested in [18], which was a superior design for emulating the virtual inertia.

1.3. Contribution

As per Table 1, several eminent controllers have been organized, where M. A. Torres [2] suggested a self-tuning virtual synchronous machine that tunes the inertia constant as per the severity of disturbance. Further, J. Alipoor demonstrated the concept of alternating the moment of inertia in [3]; however, it only tunes the constants (inertia and damping) with a slothful response. M. H. Ravanji et al. suggested a swing equation-based virtual inertia, in which the DFIG wind turbines participate in frequency regulation, although it addresses only frequency oscillations and RoCoF. In [13], a fuzzy-based controller has

been suggested, but it exhibits a sluggish response after the disturbance. A. Karimi et al. in [17] presents a fuzzy-based controller and computes the correction term needed to adjust the governor output power during a disturbance; however, as in [13], the same controller suffers from a sluggish response. As per the power voltage characteristics, renewable generation is a non-linear system, and consequently, non-linear controllers have been introduced in the literature. In addition to this, artificial intelligence (AI)-based VSG is projected in [13]; however, non-linear controllers need accurate parameter identification for a superior performance; therefore, a model prediction controller [18] has been further explored to determine the optimal power requirement by RES during the disturbance. It determines the real power for IIDG in real-time, using a predictive framework. However, the performance evaluation of the suggested MPC suffers in the identification of the optimal real power set point. Primarily, the proposed work focuses on a comparison with the eminent state of the art, as discussed above, and reveals the performance of the proposed controller. Conventional constant parameter (CP-VSG) and zero VI have also been taken into consideration in numerous case studies. A time-domain analysis obtained on an IEEE 14 bus test system developed in MATLAB, and the simulation results, corroborate the superiority of a proposed STO-based STC over the current state of the art. Time-domain specifications such as settling time, DC voltage variation, and generator-1 settling time have been exploited for numerous popular schemes viz. fuzzy-based VSG, AI-based VSG, and MPC schemes subjected to three phase faults on bus-12. These comparisons have been compiled in later sections. Furthermore, the normalized power supplied during the disturbance has also been plotted for various schemes, as discussed above, and compared with this STO-STC scheme. The super-twisting observer can work in the presence of any kind of bounded disturbances and converge in finite time. Further, if one state of the system is known, then also it can track all the information of the system in finite time. Thus, it reduces the required number of sensors in the system.

Table 1. Comparison of current work eminence and state of the art.

Existing State of the Art	Controller Description	Attributes
M. A. Torres et al., 2014 [2]	Self-tuning virtual synchronous machine	Energy storage system for inertia emulation
J. Alipoor et al., 2015 [3]	Alternating inertia-based virtual SG	Tunes inertia and damping constants
M. H. Ravanji et al., 2017 [4]	Swing equation-based virtual inertia	DFIG-based wind turbines
C. A. B Karim et al., 2018 [13]	VSG based on fuzzy controller	Distributed generation in microgrid
S. Wang et al., 2019 [15]	Advanced control solutions	Enhanced resilience in distribution system
A. Karimi et al., 2020 [17]	Fuzzy-based VSM	Regulated governor output and correction term
A. A-Idowu et al., 2021 [16]	MPC (model predictive control)	Optimal power set points
Proposed work	STC based on STO	Enhance frequency dynamics in multi-machine system

The key outcomes of this work have been given below:

1. Super-twisting is a novel scheme to identify all the states in finite time $t < T_0$ with the minimum number of required sensors.
2. The super-twisting controller converges very quickly and provides the accurate real power required for inertia emulation under finite disturbance $d1$.
3. This novel controller efficiently improves the frequency dynamics.
4. The suggested controller illustrates a superior performance over the recent popular controllers viz. model prediction, fuzzy, or self-tuning controllers.

2. Mathematical Modeling of Proposed VSG Dynamics

In this section, the VSG-IIDG dynamic equation has been developed in the d-q reference frame [19], which is based on two control loops corresponding to the active power and reactive power exchange.

2.1. Non-Linear Dynamics of VSG-IIDG

The 3- Φ active power is basically computed as [20,21], which is given below:

$$P_e = v_d i_d + v_q i_q \quad (1)$$

whereas v_d, v_q, i_d, i_q are voltages and currents in inverter terminals on the d-q reference frame. The transformation from a-b-c to d-q requires the phase angle θ , which is computed from the second-order generalized integrator, SOGI PLL [22]. In a conventional synchronous machine (SM), a well-known inertia equation viz. swing equation of rotor dynamics has been given as follows.

$$J\omega_m \dot{\omega}_m = P_m - P_e - D(\omega_0 - \omega_m) \quad (2)$$

where, ω_0 is the nominal angular frequency of VSG-IIDG, ω_m is the measured angular frequency obtained through SOGI-PLL, J is the moment of inertia, and D is the damping coefficient [23]. Further, a high value of D brings the measured frequency more quickly to the nominal value. P_m is the governor output power. Further, as with conventional SM, the governor model is included in this modeling, as given in Equation (3):

$$P_m = P_0 + K_g(\omega_0 - \omega') \quad (3)$$

where, P_0 is the set active power and K_g is the governor droop coefficient. ω' is the angular frequency generated by the swing equation. Furthermore, one more loop corresponds to the reactive power Q_e for voltage droop control. The reactive power on the d-q coordinates are given as [20,21]:

$$Q_e = v_d i_q - v_q i_d \quad (4)$$

The output voltage of the VSG-IIDG is governed by the following dynamics:

$$V = (K_p + \frac{K_I}{s})(Q^* - Q_e) \quad (5)$$

K_p, K_I are PI (proportional and integral gains). Q^* is the reactive power reference. In addition, the voltage droop generates reference Q^* , which is given by:

$$Q^* = Q_0 + K_q(V_0 - V_r) \quad (6)$$

where K_q is the droop coefficient and Q_0 is the set reactive power. Further, V_r is the rms value of the inverter side voltage (PCC), and V_0 is the reference voltage. Now, upon differentiating Equation (5), we get:

$$\dot{V} = K_p \dot{Q}^* + K_I Q^* - K_p \dot{Q}_e - K_I Q_e$$

Thus,

$$\dot{V} = K_p \dot{Q}^* + K_I Q^* - K_p \dot{Q}_e - K_I Q_e \quad (7)$$

\dot{Q}^* can be neglected with respect to the frequency dynamics; therefore:

$$\dot{V} = -K_p \dot{Q}_e + K_I(Q^* - Q_e) \quad (8)$$

Now, Equations (2) to (8) illustrate the non-linear behavior of VSG-IIDG. Therefore, a non-linear controller such as the sliding mode control (SMC) is required to handle these

(TV) disturbances within finite time $t < T_0$. STO is based on a high-order SMC (HOSMC). The STO dynamics [26,27] for the estimation of the system states (a copy of the dynamic system) have been evaluated as:

$$\left. \begin{aligned} \dot{\hat{x}}_1 &= \zeta_1 + \hat{x}_2 \\ \dot{\hat{x}}_2 &= \zeta_2 + u \end{aligned} \right\} \quad (10)$$

In STO design, we neglected the unknown d_1 ; further, ζ_1, ζ_2 are correction terms. The error variables are defined as $e_1 = x_1 - \hat{x}_1$ and $e_2 = x_2 - \hat{x}_2$.

Our objective is to estimate x_1, x_2 in finite time, $t < T_0$. Furthermore,

$$x = \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} = \begin{bmatrix} \omega_m \\ \dot{\omega}_m \end{bmatrix} \quad (11)$$

Now, the derivative of the error, $\dot{e}_1 = \dot{x}_1 - \dot{\hat{x}}_1 = x_2 - \hat{x}_2 - \zeta_1 = e_2 - \zeta_1$

Similarly, $\dot{e}_2 = \dot{x}_2 - \dot{\hat{x}}_2 = u + d_1 - u - \zeta_2 = d_1 - \zeta_2$

Selecting ζ_1, ζ_2 in such a way that $e_1, e_2 \rightarrow 0 \forall t \leq T_0$; thus, as per [25], the correction terms will be as given as $\zeta_1 = k_1 |e_1|^{\frac{1}{2}} \text{sign}(e_1)$ and $\zeta_2 = k_2 \text{sign}(e_1)$. In these equations, e_2 is missing due to the lack of information regarding this error variable, and further, the correction factors are based on e_1 only. Now, STA based on e_1, e_2 is given as follows:

$$\left. \begin{aligned} \dot{e}_1 &= -k_1 |e_1|^{\frac{1}{2}} \text{sign}(e_1) + e_2 \\ \dot{e}_2 &= -k_2 \text{sign}(e_1) + d_1 \end{aligned} \right\} \quad (12)$$

It has been assumed that the finite disturbance $|d_1| < \delta_0$. Now, based on the literature available [19,25], if we choose $k_1 = 1.5\sqrt{\delta_0}$ and $k_2 = 1.1\delta_0$, which have assured convergence, $e_1, e_2 \rightarrow 0 \forall t \leq T_0$, and finally:

$$\left. \begin{aligned} x_1 &= \hat{x}_1 \\ x_2 &= \hat{x}_2 \end{aligned} \right\} \forall t \leq T_0 \quad (13)$$

Therefore, in a second-order system, using a higher-order sliding mode observer viz. STO, one can estimate numerous states of the system in finite. This observer exposes an acceptable performance in terms of finite time robust estimation. Now, the next section discusses the controller design.

2.4. ST Control Algorithm and Design

For the controller design, a relative degree of two is required. However, STC would be applicable for a relative degree of one only. Accordingly, the sliding manifold has been modified to obtain a required relative degree of one.

$$\left. \begin{aligned} S &= cx_1 + x_2 \\ \dot{S} &= c\dot{x}_1 + x_2 = cx_2 + u + d_1 \end{aligned} \right\} \quad (14)$$

Let the controller be given as $u = -cx_2 + \psi(t)$ as shown in Figure 2, where, according to the ST algorithm, $\psi(t)$ is defined as:

$$\psi(t) = -\lambda_1 |S|^{\frac{1}{2}} \text{sign}(S) - \lambda_2 \int_0^t \text{sign}(S) d\tau \quad (15)$$

where $\lambda_1 = 1.5\sqrt{\delta_0}$ and $\lambda_2 = 1.1\sqrt{\delta_0}$. Further, Δ is defined on the finite disturbance as $\delta_0 \geq |d_1|$. After the control input, we get:

$$\left. \begin{aligned} \dot{x}_1 &= S - cx_1 + e_2 \\ S &= ce_2 - \lambda_1 |S|^{\frac{1}{2}} \text{sign}(S) - \lambda_2 \int_0^t \text{sign}(S) d\tau + k_2 \text{sign}(e_1) \end{aligned} \right\} \quad (16)$$

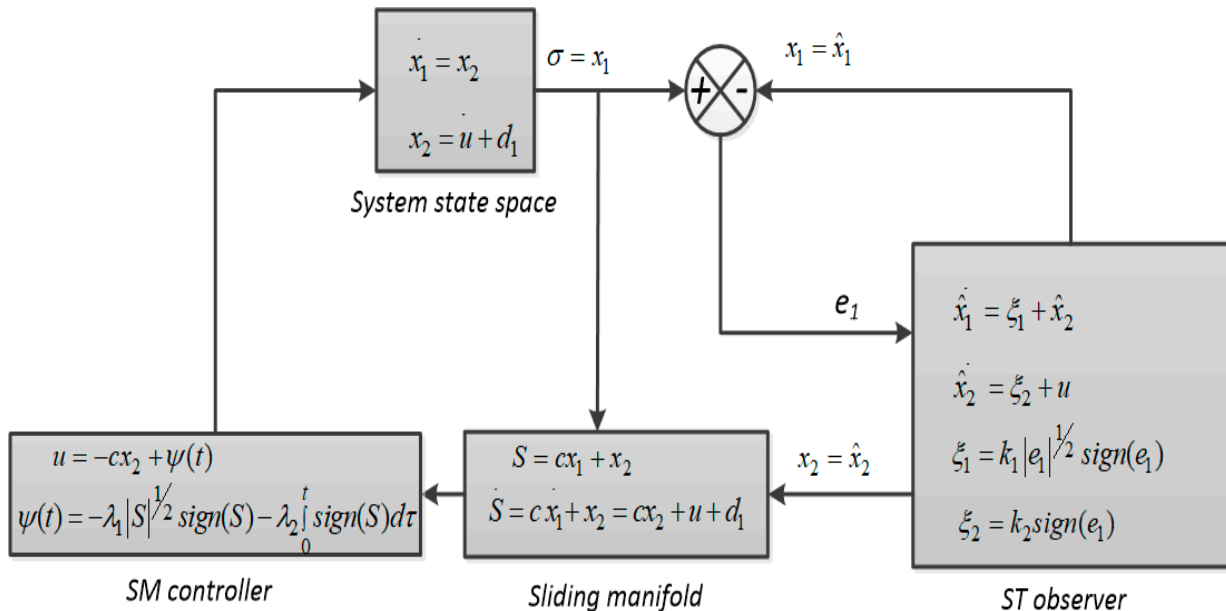


Figure 2. Sliding mode observer-based sliding mode controller.

Furthermore, the Lyapunov stability of the suggested STO-based STC has been dependent upon the error dynamics e_1, e_2 . Since the correction factors are based on e_1 only, the Lyapunov function has been chosen for σ . For STC, the scaled function has been taken as $\dot{\sigma} = -k_1 |\sigma|^{\frac{1}{2}} \text{sign}(\sigma)$. Now, assuming the Lyapunov function as follows:

$$\|V\| = \frac{1}{2} \sigma^2, \sigma \in \mathfrak{R} \quad (17)$$

V is continuous and positively defined ($V > 0$). However, it is non-differentiable at $V(0) = 0$. The error dynamics should be globally asymptotically stable, based on the selected ST gains.

Now,

$$\|V\| = \sigma \dot{\sigma} = -k_1 |\sigma|^{\frac{1}{2}} \sigma \text{sign}(\sigma) = -k_1 |\sigma|^{\frac{3}{2}} < 0 \quad (18)$$

where $\sigma \text{sign}(\sigma) = |\sigma|$. Thus, it has been concluded that the error dynamics converge to zero in finite time $t \leq T_0$. To obtain the effectiveness of the STO-based STC controller, here, we tried to observe and explore the attributes of the suggested controller. Consider a very simple physical system that is a moving car system driven by some controlled input (u). The system state space can be written based on Newton's law, and the displacement vector is given as x_1 and the velocity vector $\dot{x}_1 = x_2$. Further, at the equilibrium point, $\dot{x}_1 = x_2 = 0$. This simple system has been examined with the proposed controller, where the disturbance is $d_1 = 0.5 \sin \omega t$ with $\omega = 1$ rad/s. The sliding surface has been chosen as $S = x_1 + \frac{1}{357.56} x_2$. For simulation purposes, the controller gains have been given as follows.

- STO gains: $k_1 = 2.1, k_2 = 2.2$;
- STC gains: $\lambda_1 = 2.1, \lambda_2 = 1.55$;
- Constant $c = 1$, sampling time (MATLAB) = 1 ms.

The twisting observer and controller algorithm have been written in a MATLAB function, and the prime objective of the controller is to bring the vehicle to the desired position, starting from the zero position. The simulation results have been portrayed in Figure 3. As per the results obtained in Figure 3, the suggested controller works well with a good precision of 10^{-3} .

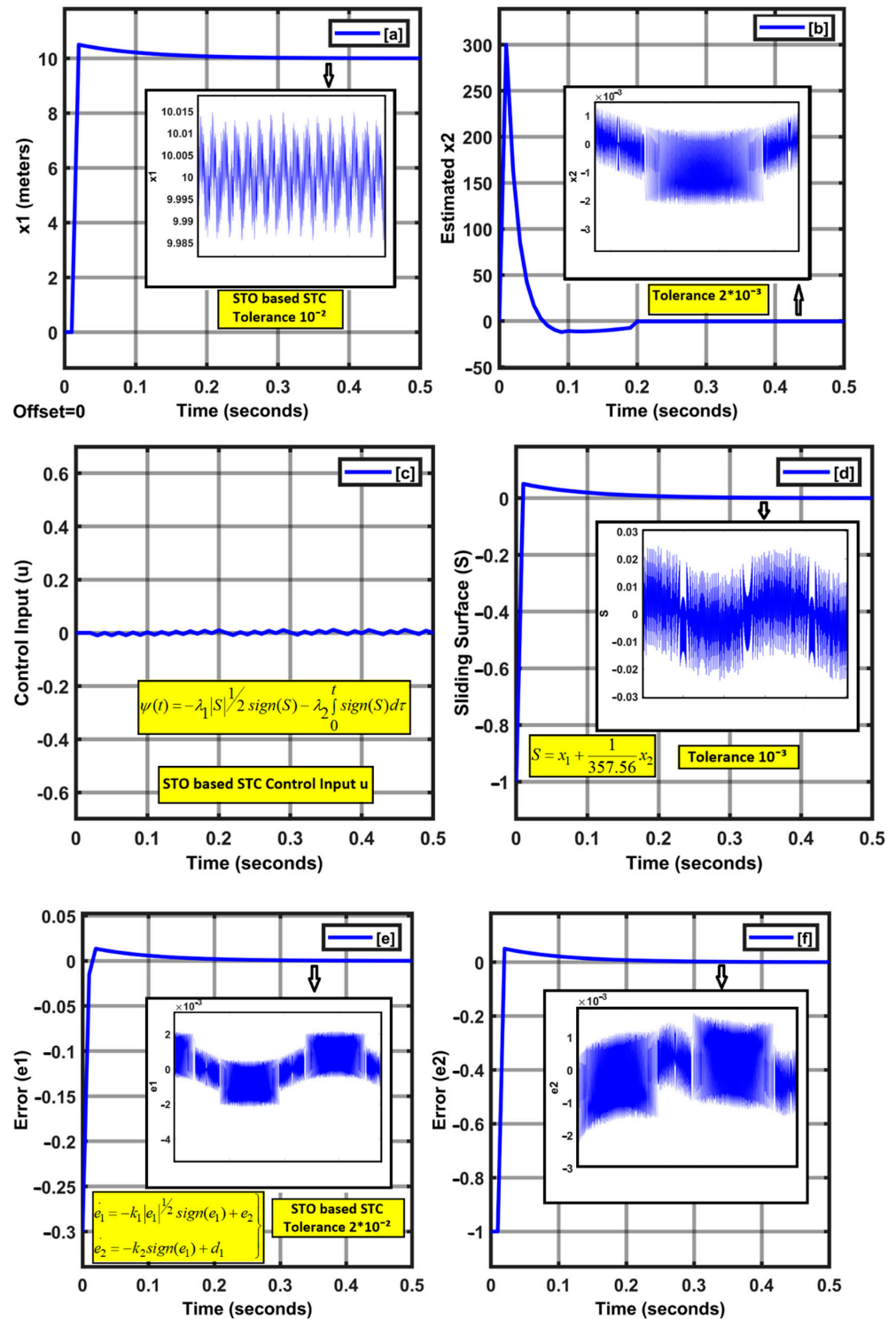


Figure 3. (a) Position vector x_1 evolution, (b) estimated state vector x_2 evolution, (c) control input (u) evolution, (d) sliding surface (S) evolution, (e) observer error (e_1) of state x_1 , (f) observer error (e_2) of state x_2 .

Furthermore, the same controller has been utilized for virtual inertia emulation on the IEEE 14 bus test system, with the output $\sigma = \Delta P$, where ΔP is the additional power required for the virtual inertia emulation, as per the frequency deviation $\Delta\omega_m$ and the rate of change of frequency $\dot{\omega}_m$.

In this paper, a sudden load change has been taken as 2 pu. Therefore, $\delta_0 = 2$ has been selected. The STO and STC gains have been estimated earlier in this section.

3. Test System Configuration and Setup

In this paper, the IEEE 14 bus system has been taken [23] as a test system. A 415 V photovoltaic system (IIDG) 0.5 MW has been incorporated at bus-12 through a 15 MVA transformer. Several synchronous generators are connected at various buses in isolation, or with some local demands.

Furthermore, various static and dynamic loads have been connected at different buses, as shown in Figure 4. At bus-9, the static Var compensator of 10 MVAR has been coupled. A 50 MW wind turbine generator (WTG) is connected at bus-14 to further supply a high demand at bus-13. In the test system, to obtain sufficient information on the voltage, power flow and angle, power flow analysis has been conducted, and the whole test system has been designed and developed on a MATLAB/Simulink MathWorks® platform. The open-circuit voltage of PV generation is 800 V, with a 600 A short circuit current. The fuses are usually incorporated to protect the PV generation, as shown in Figure 4, and are incorporated at various places. Now, for the inertia emulation, MPPT (maximum power point tracking) does not play a significant role, and it is omitted in the development of the MATLAB detailed model of a test system. The case study has been executed based on various factors, viz. sudden load change and intentional islanding. Further, during simulation, the temperature or irradiance variation have been omitted for simplicity, and a detailed model has been simulated based on ode 23tb, with a maximum step size of 1 ms. The voltage source inverter is an IGBT-based 2 level bridge inverter, which is efficiently regulated through pulse modulation. In this work, STO-based STC accurately measures ΔP based on the nominal frequency deviation and the rate of change of frequency $(\omega_m, \dot{\omega}_m)$, owing to the sudden bounded disturbance d_1 . The super-twisting observer and the controller algorithm have been already given in Equations (9)–(18), and consequently, the code is written in a MATLAB function. The efficacy of the proposed controller has been verified by numerous simulation results. The system parameters used in building the test system in MATLAB have been organized in Table 2.

Table 2. Test system parameters.

System Parameters	Values
DC-link voltage (V_{DC})	800 V
Nominal frequency (f_n)	50 Hz
Output inverter voltage	450 V
Inertia constant (H)	4, 7, 10 s
Maximum irradiance	1000 W/m ²
Temperature (T)	38 °C
Parallel strings	76
Series modules	23
Open circuit voltage (V_{oc})	36.3 V/Module
Short circuit current (I_{sc})	7.84 A/Module
Nominal voltage	835 V(DC)
Nominal current	598 A(DC)
Max power/module	213.15 W
Speed regulation of governor (R)	0.05 pu
Inertia of synchronous generator (H)	4 s

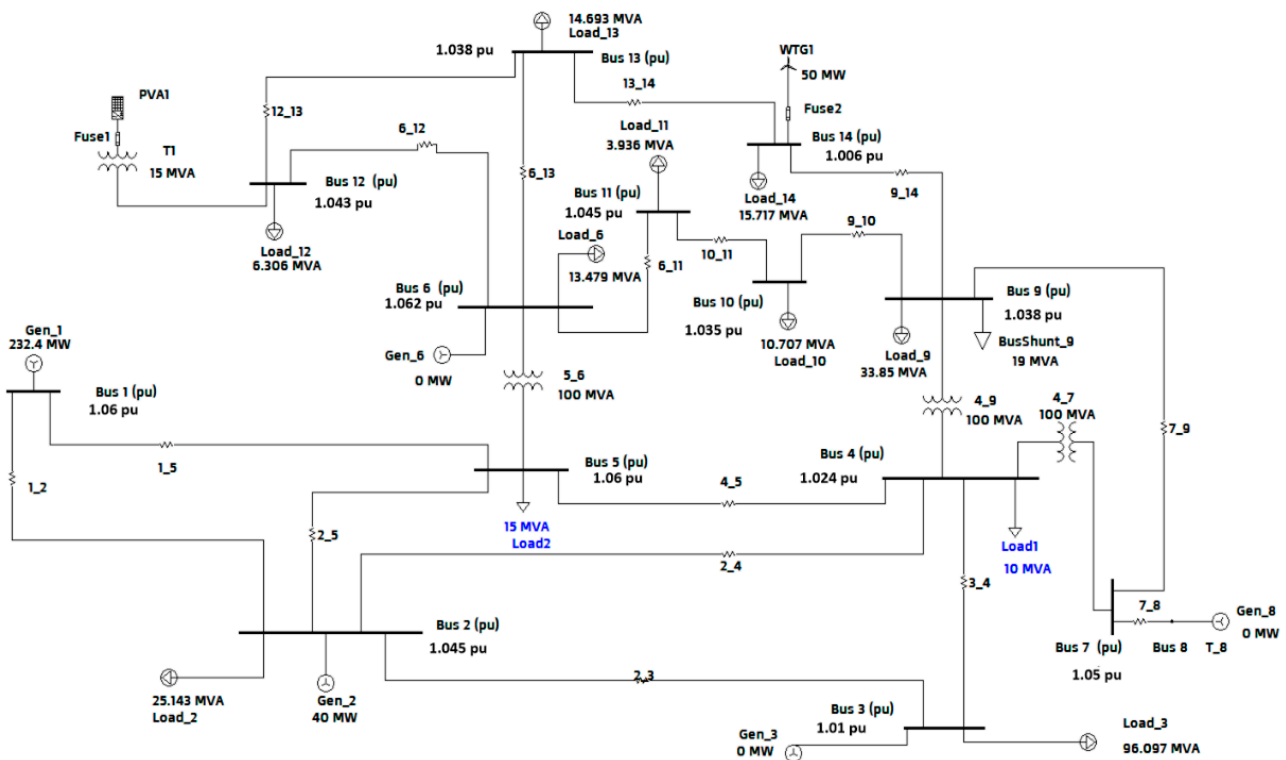


Figure 4. IEEE 14 bus test system (bus voltages in pu after power flow analysis).

4. Simulation Results and Performance Evaluation

As per the available literature [23–25], four main consecutive control steps have been utilized to restore the deviated frequency to the nominal value. 1. Inertial control 2. Primary control 3. Secondary control 4. Tertiary control. The fastest control is generally exhibited by SG, which instantaneously responds to supply–demand disturbance before the primary controller is activated. However, in recent years, SGs have been replaced by IIDGs to a large extent, and thus, the overall system damping and inertia are reduced significantly. Therefore, during load throw-off or numerous kinds of faults, primary control is not an effective solution to many problems, due to their sluggish response, which in the worst conditions causes blackouts or a complete failure of the system. The above discussion is shown in Figure 5.

4.1. Sudden Load Variation in Grid-Connected VSG

In this scenario, the capability of the proposed STO-based STC has been investigated. The load changes at bus-12 from 500 kW (for simulation purposes, the load connected at bus-12 was taken to 500 kW instead of 6.306 MVA for simplicity in designing the MATLAB model) to 650 kW at $t = 2$ s, and subsequently, the frequency droop has been observed from the nominal value f_n ; consequently, a load has been decreased by up to 100 kW at bus-12 at $t = 15$ s (a sudden load variation through variable AC resistors in MATLAB) and a frequency overshoot, along with RoCoF, has been investigated for the selected state of the arts.

The frequency dynamic for different control schemes has been portrayed in Figure 6. An IIDG with zero or negligible inertia suffers from an impulsive overshoot; nonetheless, the oscillating time has been drastically reduced, as compared to the constant parameter, viz. the CP-VSG scheme (a constant inertia fixed up to $H = 4.5$ s).

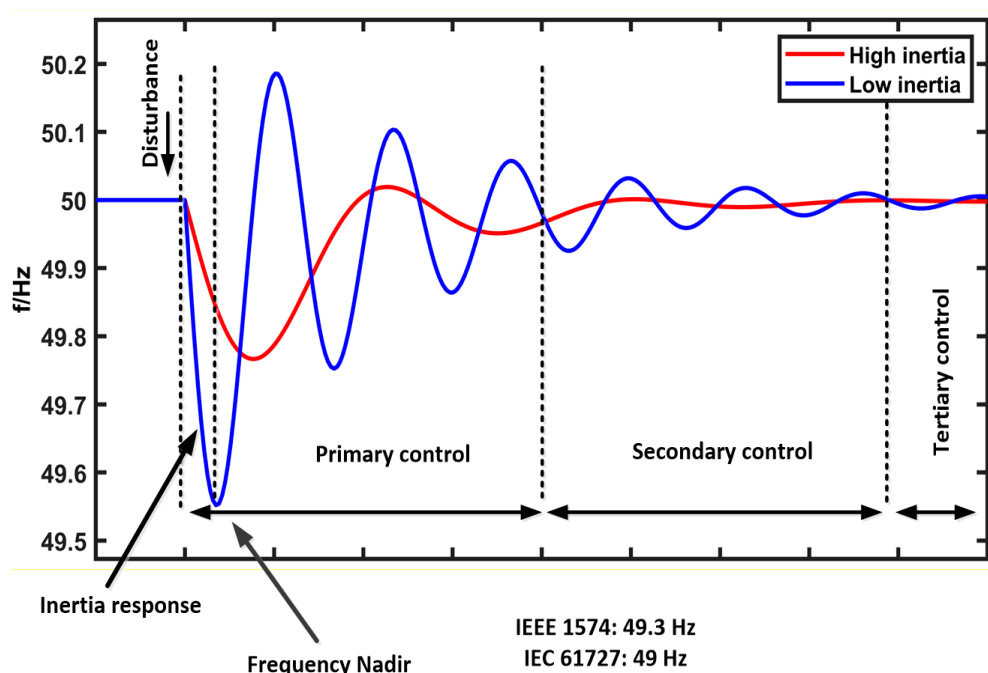


Figure 5. Frequency dynamics over time with various control steps.

The obvious disadvantages of a fixed high-inertia system viz. CP-VSG are sustained periodic oscillations and a subsequently inferior dynamic performance. Furthermore, fuzzy-based VSG illustrates a superior inertia emulation during the disturbances, but the settling time of oscillation is high, as compared to the proposed STO-based STC. The proposed STO-based STC demonstrates RoCoF = -0.25 Hz/s at t = 2 s, and 0.75 Hz/s at t = 15 s, with a 49.9 Hz frequency nadir and a 50.5 Hz frequency overshoot. The investigated results recommend a superior dynamic performance of STO-based STC over numerous control schemes, as shown in Figure 6. The inertia emulation for various schemes has been examined, based on the swing equation [23], and is given as:

$$J = \frac{\Delta P_m - \Delta P_e - D(\omega_m - \omega_0)}{\dot{\omega}_m} \tag{19}$$

Now, the inertia J , and consequently, the inertia constant $H(s)$ adaptively increased during the disturbances. STO-based STC efficiently emulates and regulates the inertia during disturbance, which is constant in the CP-VSM scheme. An encapsulation of the time-domain performance evaluation of popular controllers and STO-based STC on sudden load variation has been compiled in Table 3.

Table 3. Performance evaluation and comparison on sudden load change.

Attributes	Control Scheme	No VI	CP-VSG	STO-Based STC
Frequency nadir [Hz]		49.2 (Violates IEEE 1574)	49.4	49.8
Frequency overshoot [Hz]		52 (Violates IEEE 1574)	51.9	50.9
RoCoF [Hz/s]		-0.8	-0.7	-0.3
Steady state error (%)		1.45	0.7	0.13
Settling time [t _s /s]		5	10	3
Inertia H [s]		0	4.5	Adaptive (4.4–4.9)
Attributes	Control Scheme	Fuzzy-Based VSG	STALG	MPC Scheme
Frequency nadir [Hz]		49.8	49.7	49.6
RoCoF [Hz/s]		-0.4	-0.5	-0.3
Settling time [t _s]		4	3.7	4.2

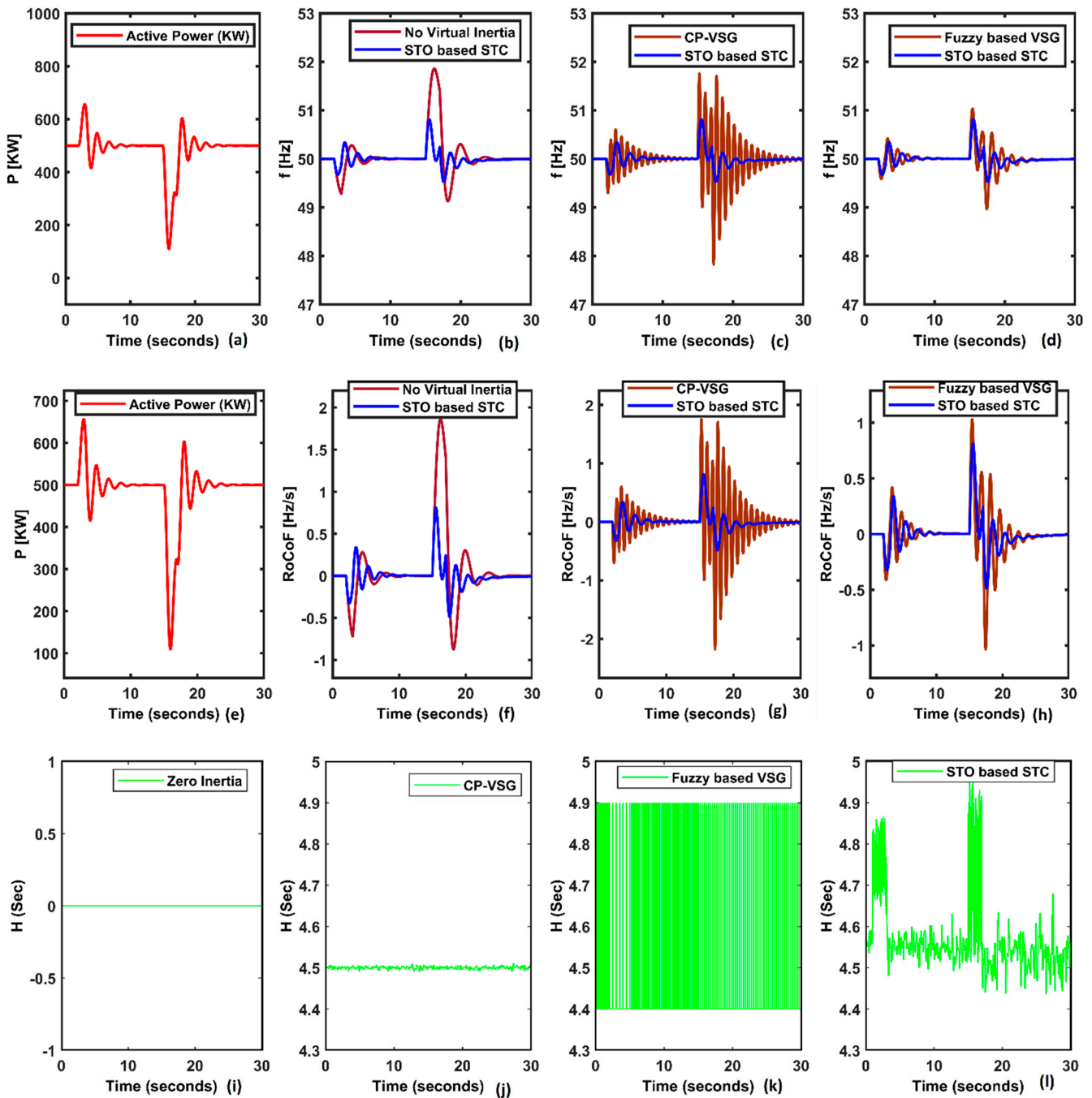


Figure 6. Grid-connected mode. (a) Active power variation, (b–d) dynamic frequency response of various control schemes, (e–h) rate of frequency variation for various control schemes, (i–l) inertia emulation.

4.2. Inertia Response on AC Fault

A 3- Φ fault has been simulated and created at 1 s, and subsequently cleared at 1.5 s at bus-13. Further, the bus-12 voltage falls as per fault impedance, which has been portrayed in Figure 7a. Subsequently, the bus-12 frequency has been estimated after the occurrence of the 3- Φ fault, and the frequency plots for zero inertia, CP-VSG, and STO-STC have been shown in Figure 7b. The frequency response illustrated the superior damping capability of STO-STC, with a settling time of below 5 s after fault clearance. No VI exhibited an impulsive frequency rise, due to zero inertia emulation, and consequently, no variation in the DC-link voltage, as shown in Figure 7c. Furthermore, to observe the off-shore power

fluctuation owing to fault occurrence at the same location, the bus-13 and SG-1 (bus-1) power oscillations have been examined, as shown in Figure 7d. This characterizes superior damping for the STO-STC-based IIDG at bus-12. Other schemes have also been compared and compiled in Tables 4 and 5, respectively.

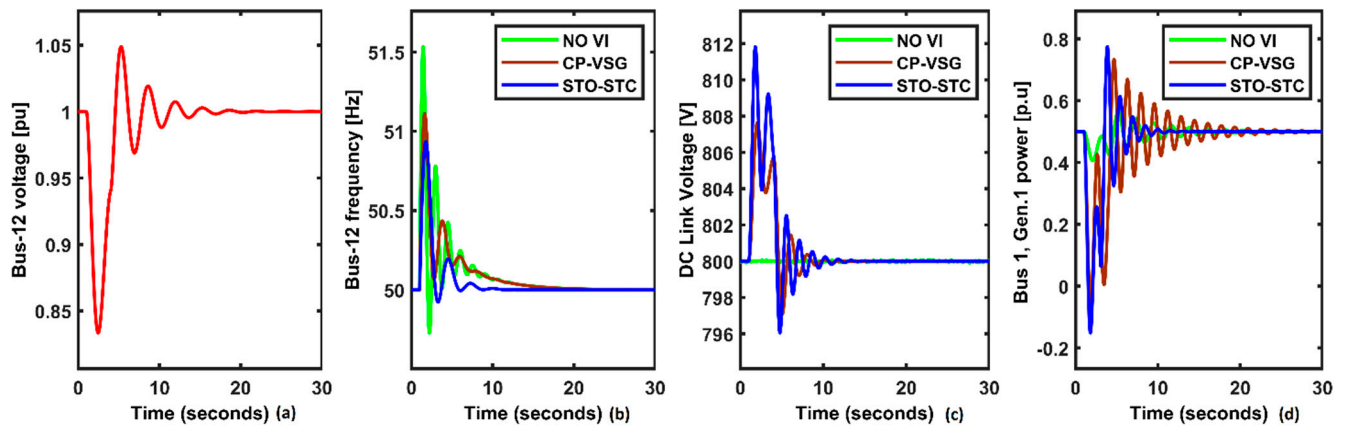


Figure 7. (a) Bus-12 voltage sag during fault (pu), (b) bus-12 frequency overshoot (Hz), (c) DC-link variation on fault (V), (d) off-shore SG power fluctuation (pu): comparison for various schemes.

Table 4. Time-domain specifications on a three phase fault (bus-12).

Attribute	Fuzzy Based VSG	STO-STC	MPC Scheme
Settling time (s)	6	6.3 oscillatory responses	5.8
DC link variation (V)	5	7	13
Bus-1, Gen.1 settling time (s)	12	9	8

Table 5. Time-domain specifications with off-shore performance assessment.

Attribute	No VI	CP-VSG	AI Based VSG
Settling time (s)	Unspecified	20 oscillatory responses	6.5
DC link variation (V)	No variation	5	9
Bus-1, Gen.1 settling time (s)	20	14	8

4.3. Normalized Active Power Response under Sudden Load Variation

Figure 8 exhibits normalized power extracted from IIDG on a sudden load change ($\Delta P_L = 50$ kW) on bus-13, next to bus-12. Numerous controllers have been taken to examine the power response at bus-12 required to mitigate the frequency variation and the frequency nadir.

It has been found that the virtual inertia response is impulsive and instantly injects active power to improve the dynamic frequency stability. Initially, IIDG was operating at 0.2 pu. Further, the injected active power dies out soon as the frequency normalizes to nominal values. Primarily, IIDG at bus-12 was delivering 0.2 normalized power, and at $t = 2$ s, sudden load changes to 50 kW and power responses have been captured in MATLAB, as shown in Figure 8. Further, in Figure 8 is a performance evaluation and comparison of various schemes on injected normalized power during a sudden load change (pu): Initially IIDG was operating at 0.2 pu.

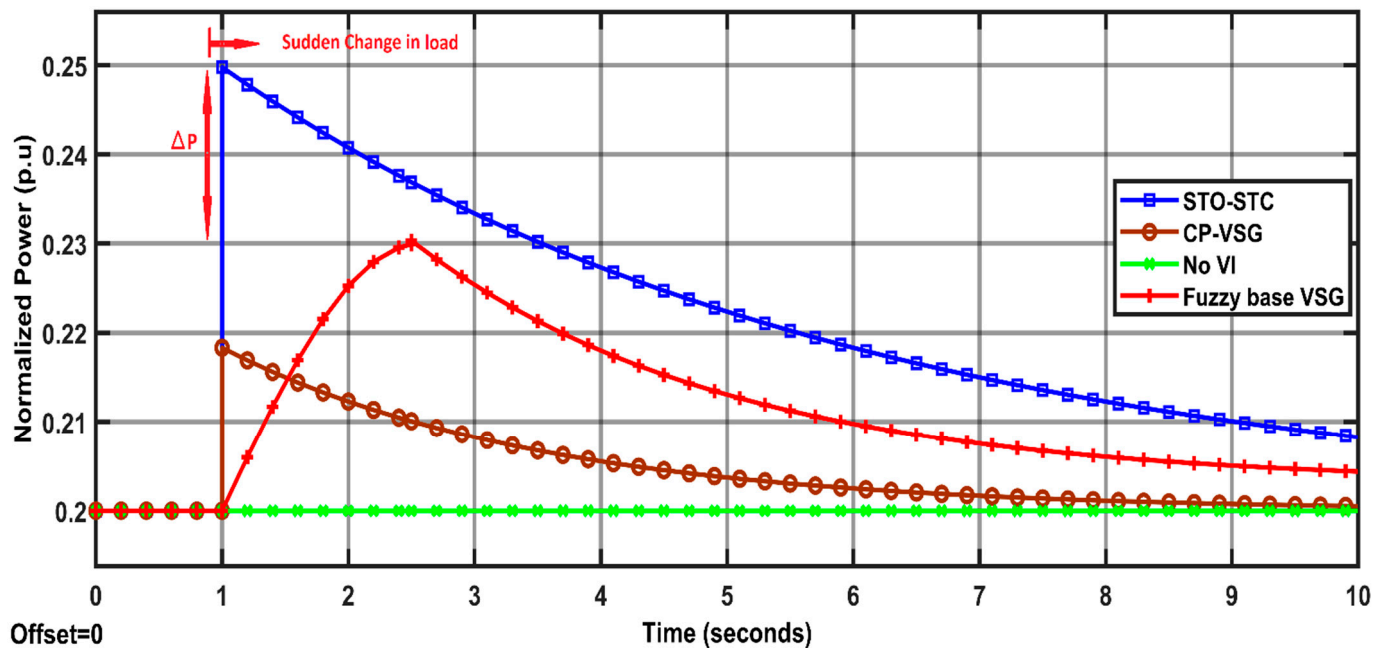


Figure 8. Performance evaluation and comparison of various schemes on injected normalized power during a sudden load change (pu).

STO-based STC estimates an accurate ΔP which is required based on the nominal frequency deviation and RoCoF ($\omega_m, \dot{\omega}_m$) owing to the sudden load change ΔP_L , which is the bounded disturbance d_1 . Further, unlike fuzzy-based VSG, STO-STC exhibits an instantaneous power response and sustains supplementary injected power until the deviated frequency has not reached the nominal frequency. The time-domain simulation results illustrate the superior performance of the STO-STC scheme over recent popular controllers viz. model prediction, fuzzy-based VSG, CP-VSG, or self-tuning controllers. It is noteworthy that the emulated inertia by the STO-STC scheme increased adaptively during the disturbance, as shown in Figure 6l. Furthermore, optimal real power ΔP has been evaluated and injected during a sudden load change, as discussed in Figure 8. The proposed STO-STC scheme observes the various states using STO, and regulates the real power flow using STC during the disturbance. This will further enhance the overall dynamic frequency stability of the system.

5. Conclusions

In this work, a novel STO-tuned STC scheme has been efficiently explored and verified using time-domain simulation results. Firstly, the inertial response has been examined in grid-connected IIDG. The virtual inertia emulation has been proven to be the most promising scheme to enhance the dynamic frequency responses of low-inertia systems. Based on the mathematical modeling of VSG, it exhibits a non-linear behavior of VSG. Therefore, a non-linear controller such as sliding mode control (SMC) is required to handle these non-linearities, which enhances the performance of VSG-IIDG. Further, STO is based on high-order SMC (HOSMC). For STC design, STO evaluates numerous system states in finite time $t < T_0$. The proposed STO-STC scheme evaluates the exact corrective active power ΔP , based on the optimal selection of system inertia, which reduces the frequency pulsations. The STO-STC scheme exhibits superior damping in a multi-machine system as compared to the popular schemes viz. the self-tuning STALG scheme, CP-VSG, fuzzy-based VSG, and the MPC scheme. This is accomplished by the optimal selection of inertia and corrective power ΔP , which regulate both the frequency and power variations. This scheme also works well in fault occurrence, which has been revealed through numerous case studies.

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